



Design and Development of an Advanced Coherent Radar Depth Sounder

Master's Thesis Defense

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Date: 16th July 2003

Committee

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Outline

- Significance of ice sheet thickness measurements
- KU radar history
- Next-Generation COherent Radar Depth Sounder (NG-CORDS)
- Advanced COherent Radar Depth Sounder (ACORDS)
- Field experiment results
- Conclusions & future work



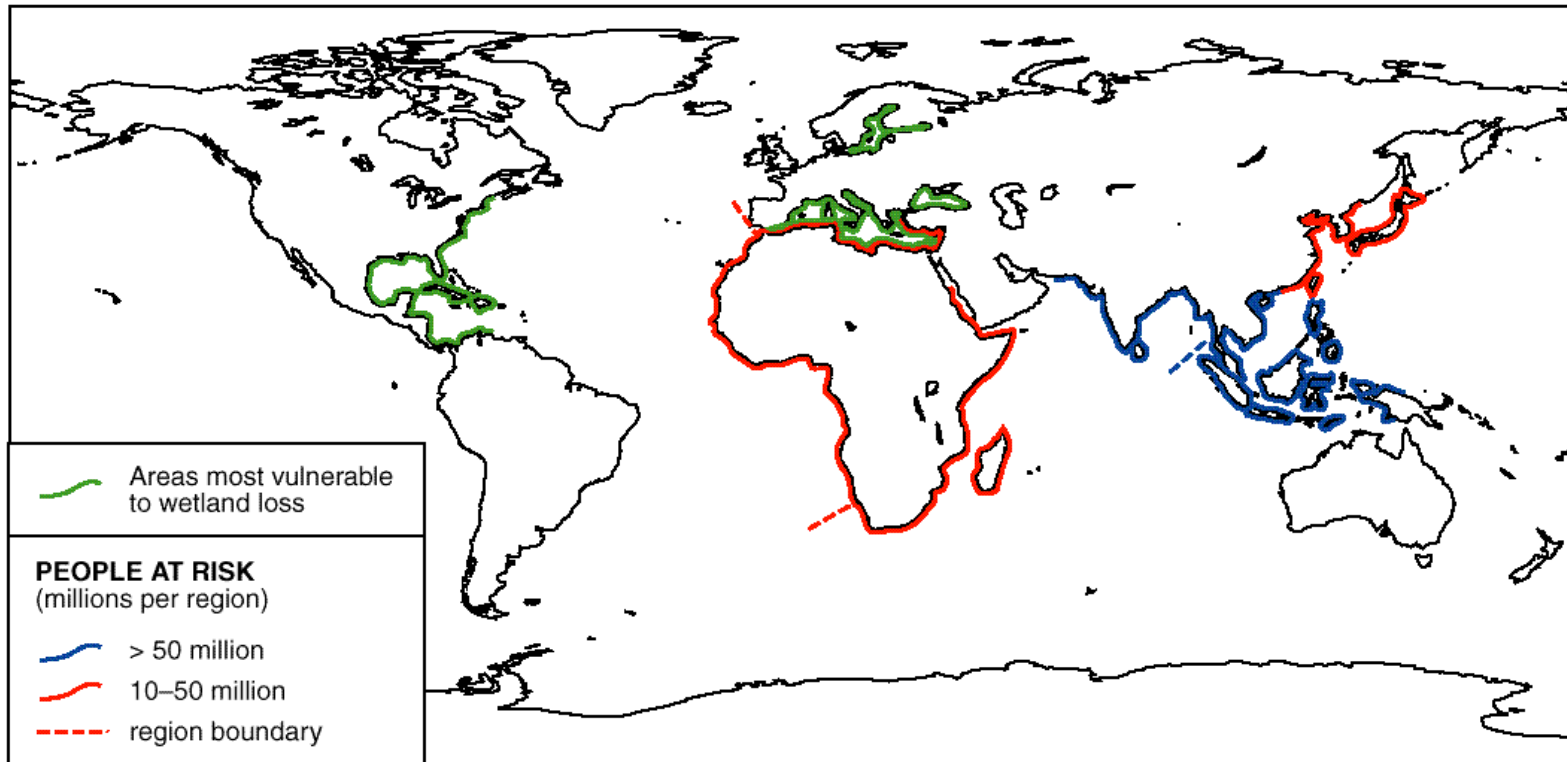
Significance of ice sheet thickness measurements

- Sea level increasing 2 mm/year
- Polar ice sheets one of the contributing factors
- Uncertainty in the contribution of polar ice sheets
- Need to assess the mass balance of ice sheets
- Ice thickness a key variable
- 1991 NASA initiative : Program for Arctic Regional Climate Assessment (PARCA)



Significance of ice sheet thickness measurements

Impact of Sea level rise on coastal regions around the world

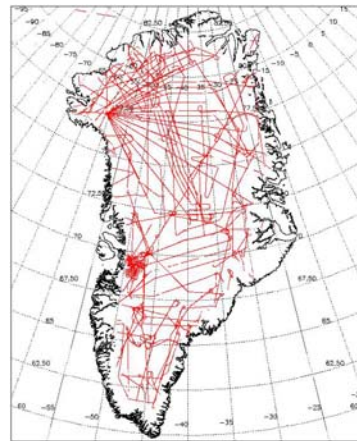


Hadley Center for Climate Prediction and Research, UK



KU radar depth sounder history

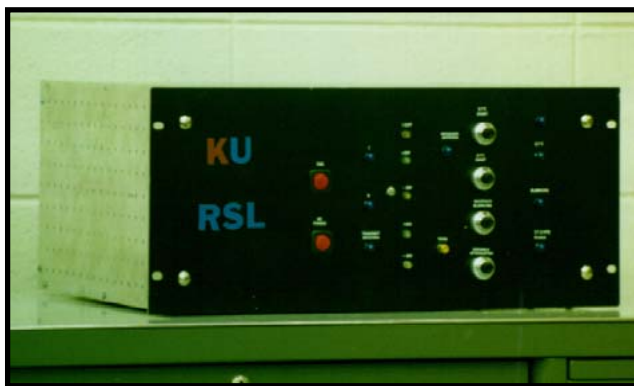
- 1980: Coherent Antarctic Radar Depth Sounder (CARDS)
- 1993: KU CARDS operated in Greenland
- 1996: Improved Coherent Antarctic and Arctic Radar Depth Sounder (ICARDS)
- 1997: NG-CORDS
- 1998: Faster digital system
- 1998-2002: 90 % success rate in field experiments





System parameters: NG-CORDS

- Surface Acoustic Wave (SAW) devices



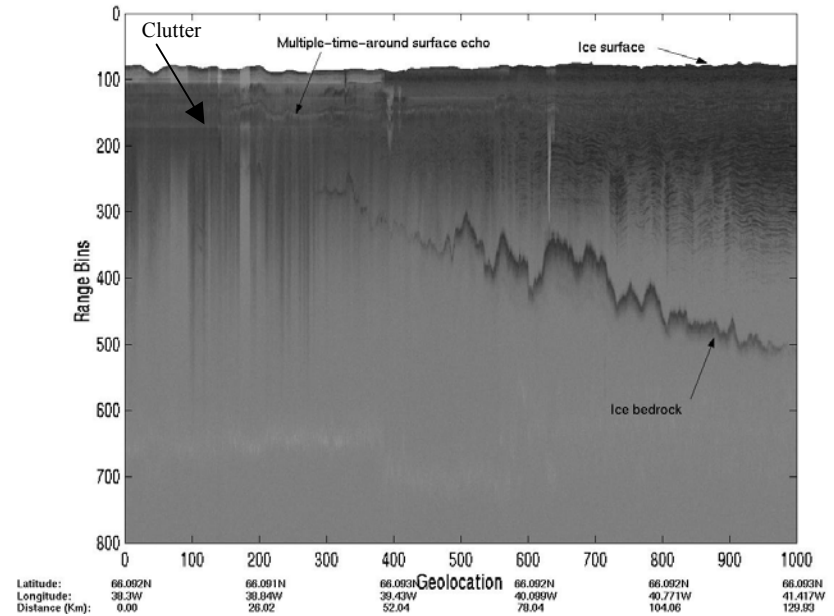
NG-CORDS analog system

Description	Characteristic	Units
Radar Type	Pulse Compression-analog	----
RF Carrier Frequency	150	MHz
RF Up-Chirp Bandwidth	17.00	MHz
Transmitted Pulse Width	1.6	μ s
Compressed Pulse Width	60	ns
Range Sidelobes	< 26	dB
Peak Transmit Power	200	W
PRF	Selectable	KHz
Number of Coherent Integrations	32 – 4,096	----
Number of Incoherent Integrations	0 – 64,000	----
A/D Dynamic Range	12-bit, 72	dB
Sampling Period	53.3 (18.75 MHz)	ns
Range Resolution	4.494	m
Antennas	4-element $\lambda/2$ dipole arrays	



NG-CORDS

- Unsuccessful in sounding few outlet glaciers & transition zones
- Off-vertical surface clutter and surface echoes mask weak basal returns
- Radio echogram from southern Greenland

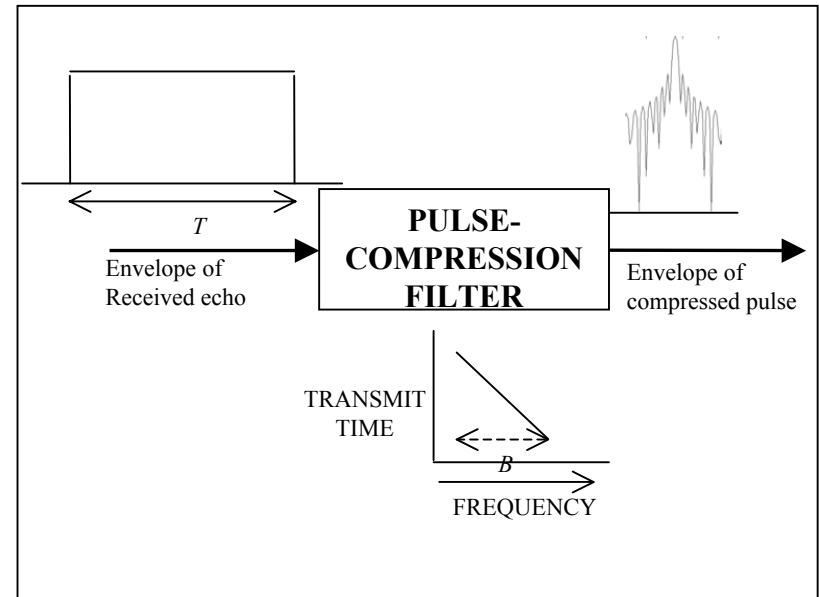




Limitations of NG-CORDS

1) Range sidelobes

- Formed due to the inherent mechanism of pulse compression
- Masks weak basal returns
- Critical in areas where ice thickness is shallow
- 26 dB sidelobes in NG-CORDS receiver

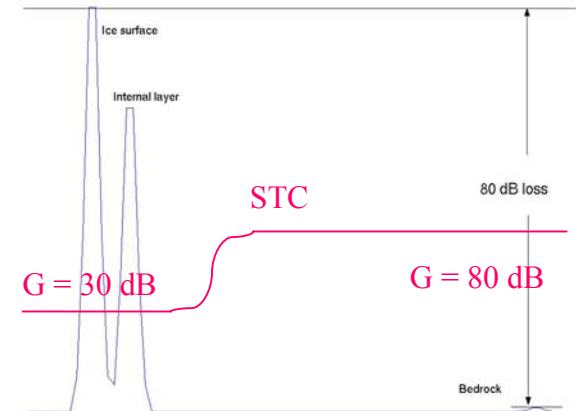




Limitations of NG-CORDS

2) Sensitivity Time Control (STC)

- Two-way loss in ice > 80 dB
- Mechanism to increase the dynamic range of the receiver
- Time varying gain
- Difficult to deconvolve the system response



3) Analog down conversion

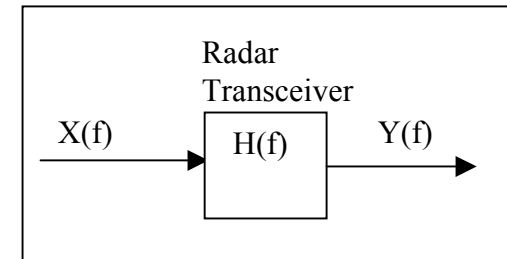
- I-Q demodulator in NG-CORDS receiver
- Amplitude or phase imbalance in the I, Q signals
- Limit the dynamic range of the receiver



Solutions for NG-CORDS limitations

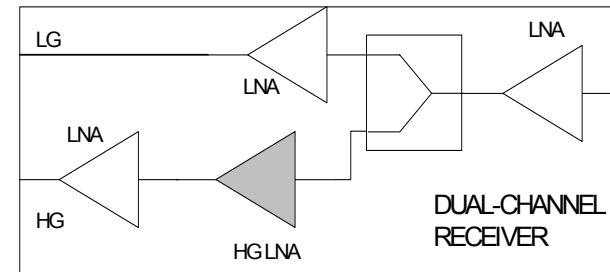
1) Reduce range sidelobes

- Mismatched filter
- $Y(f)$ is a window function
- Input $X(f) = Y(f)/H(f)$
- Needs a Waveform Generator



2) Dual-channel receiver

- Two gain channels, Low Gain (LG) and High Gain (HG)
 - No STC
- ## 3) Digital down conversion
- Bandpass sampling
 - No analog down conversion





Need for an Advanced COherent Radar Depth Sounder (ACORDS)

- Waveform Generator (WG)
- *Transmitter subsystem*
- *Dual-channel receiver with two different gain channels*
- Data ACquisition system (DAC) capable of undersampling

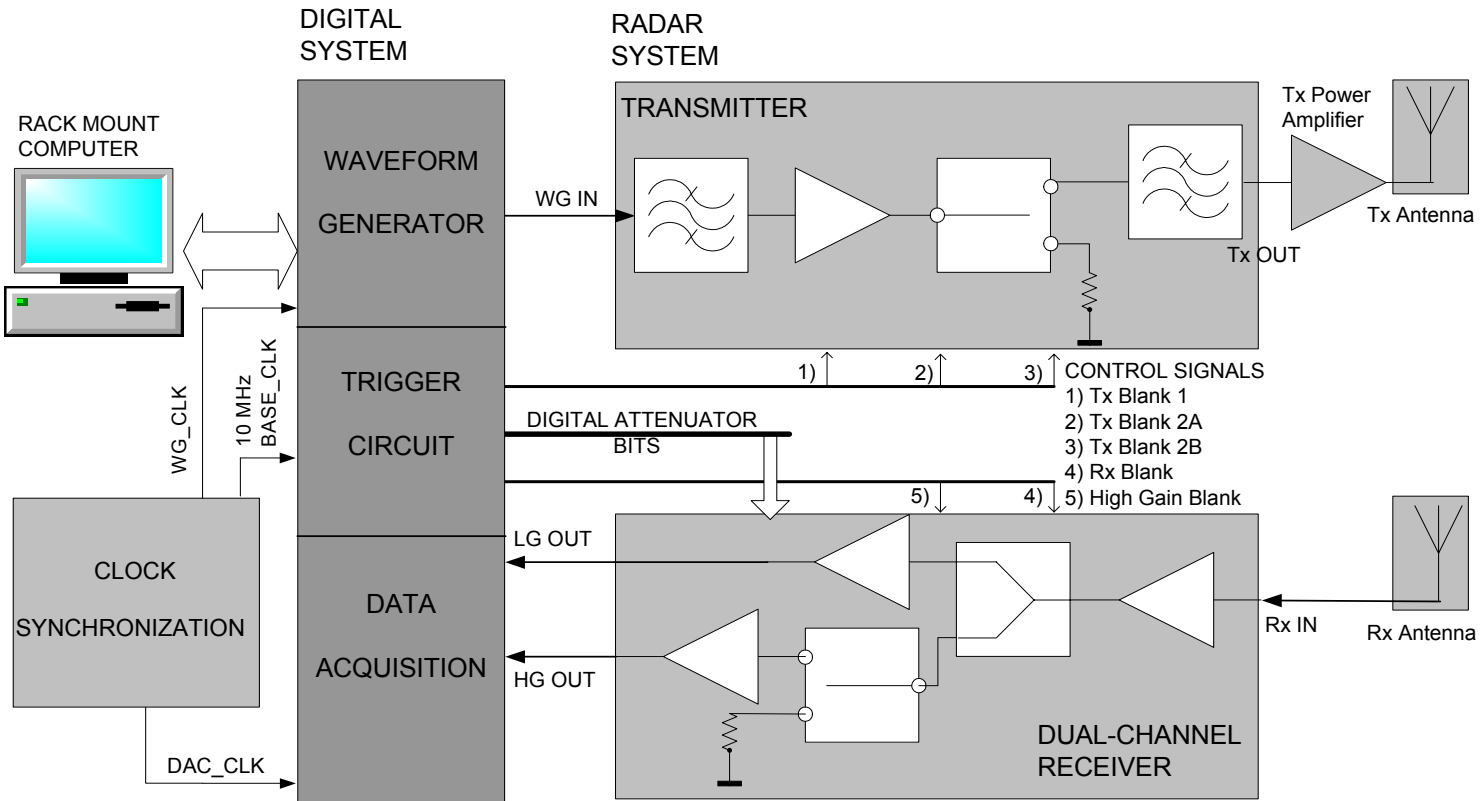


System parameters: ACORDS

Description	Characteristic	Units
Radar Type	Pulse Compression (digital)	----
RF Carrier Frequency	150	MHz
RF Chirp Bandwidth	17.00	MHz
Transmitted Pulse Width	Selectable (0.2 – 10)	μs
Range Sidelobes	< 36	dB
Peak Transmit Power	200	W
PRF	Selectable	KHz
Number of Coherent Integrations	32 – 1024	----
Number of Incoherent Integrations	0 – 64,000	----
A/D dynamic range	12-bit, 67	dB
Receiver Dynamic Range	>110	dB
Sampling Period	18.182 (55 MHz)	ns
Range Resolution	4.494	m
Antennas	4-element $\lambda/2$ dipole arrays	



ACORDS block diagram





Selection of clocks for WG & DAC: ACORDS

- Nyquist theorem requires sampling frequency to be at least 320 MHz
- Bandpass sampling allows a sampling rate at least twice the signal bandwidth
- ACORDS simultaneously oversamples and undersamples
- To avoid aliasing, sampling frequency is constrained by the following equations

$$\frac{2f_l}{k-1} \geq f_s \geq \frac{2f_u}{k}$$

$$f_s \geq 2B \text{ \& } 2 \leq k \leq \frac{f_u}{B},$$

f_s – sampling frequency

f_l – lower frequency (140 MHz)

f_u – higher frequency (160 MHz)

k – positive integer

K	Range of f_s (MHz)
2	160.00 – 280.00
3	106.67 - 140.00
4	80.00 – 93.33
5	64.00 – 70.00
6	53.33 - 56.00
7	45.71 - 46.67

Selected

f_{WG} – WG_CLK (110 MHz)

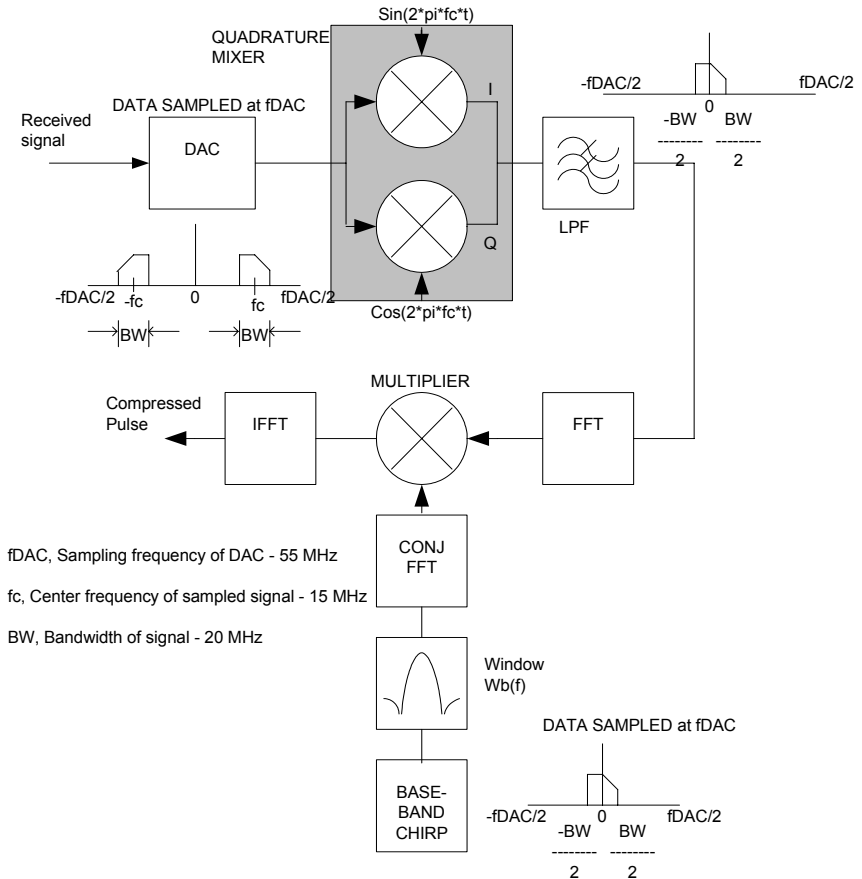
f_{DAC} – DAC_CLK (55 MHz)





Digital pulse compression: ACORDS

- NG-CORDS: SAW compressor
- ACORDS: digitally implemented
- Compression gain, $10 \cdot \log_{10}(BW \cdot T) = 16 \text{ dB}$
- Compressed pulse width, $1/BW$
- Control on frequency weighting





Derivation of receiver gain: ACORDS

HG channel gain (G_{HG})

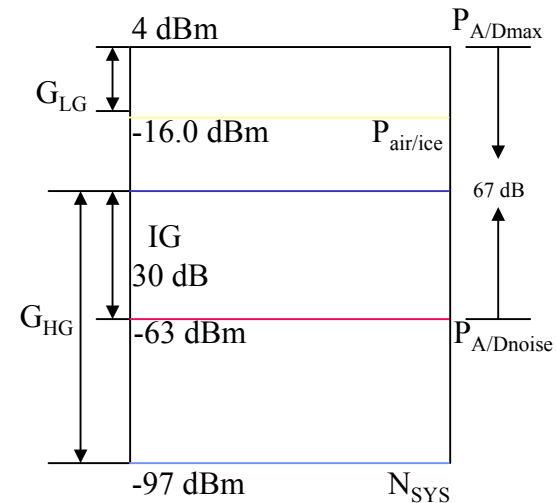
- High sensitivity
- Need at least 1000 coherent integrations
- $N_{SYS} + G_{HG} = P_{A/Dnoise} + IG$
- $G_{HG} = 64$ dB

LG channel gain (G_{LG})

- $P_{air/ice} + G_{LG} = P_{A/Dmax}$
- $G_{LG} = 20$ dB

Designed dual-channel receiver

- 80 dB gain in HG channel
- 44 dB gain in LG channel





ACORDS transmitter design

Consists of

1) Waveform Generator

- Uses Analog Devices 16-bit AD9777 as oversampling D/A converter
- First order harmonic image of the baseband chirp selected
- Generates a 140-160 MHz chirp at -7 dBm
- Transmit pulse width and window function can be programmed

2) Transmitter

Clock synchronization section

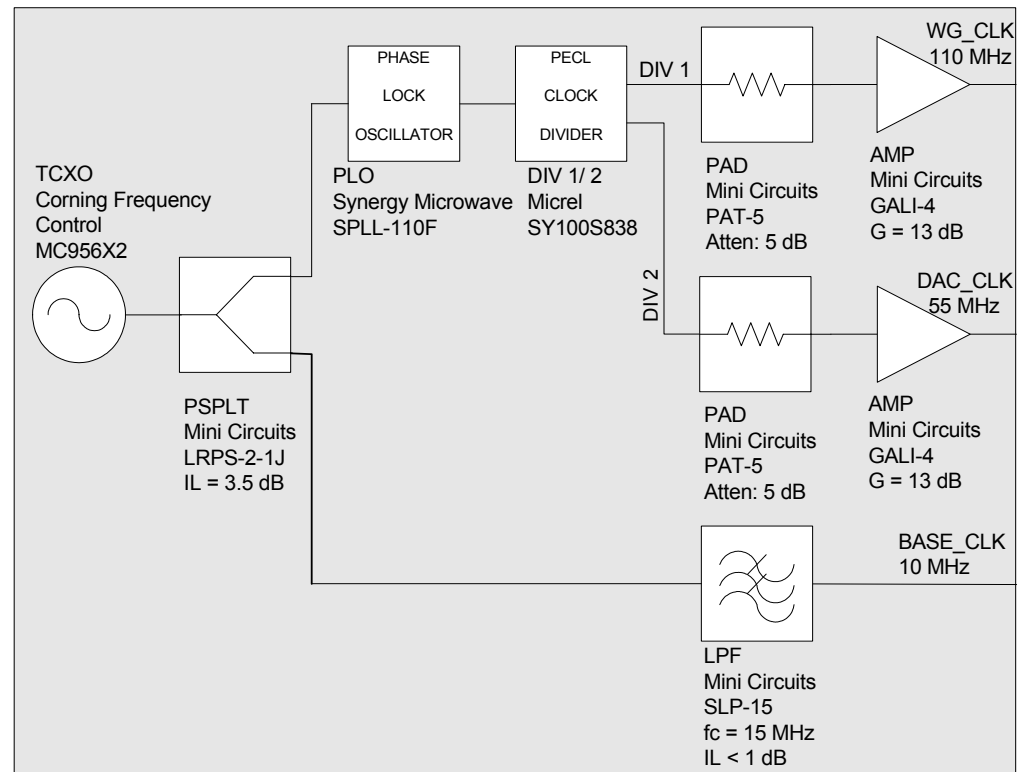
Transmitter RF section





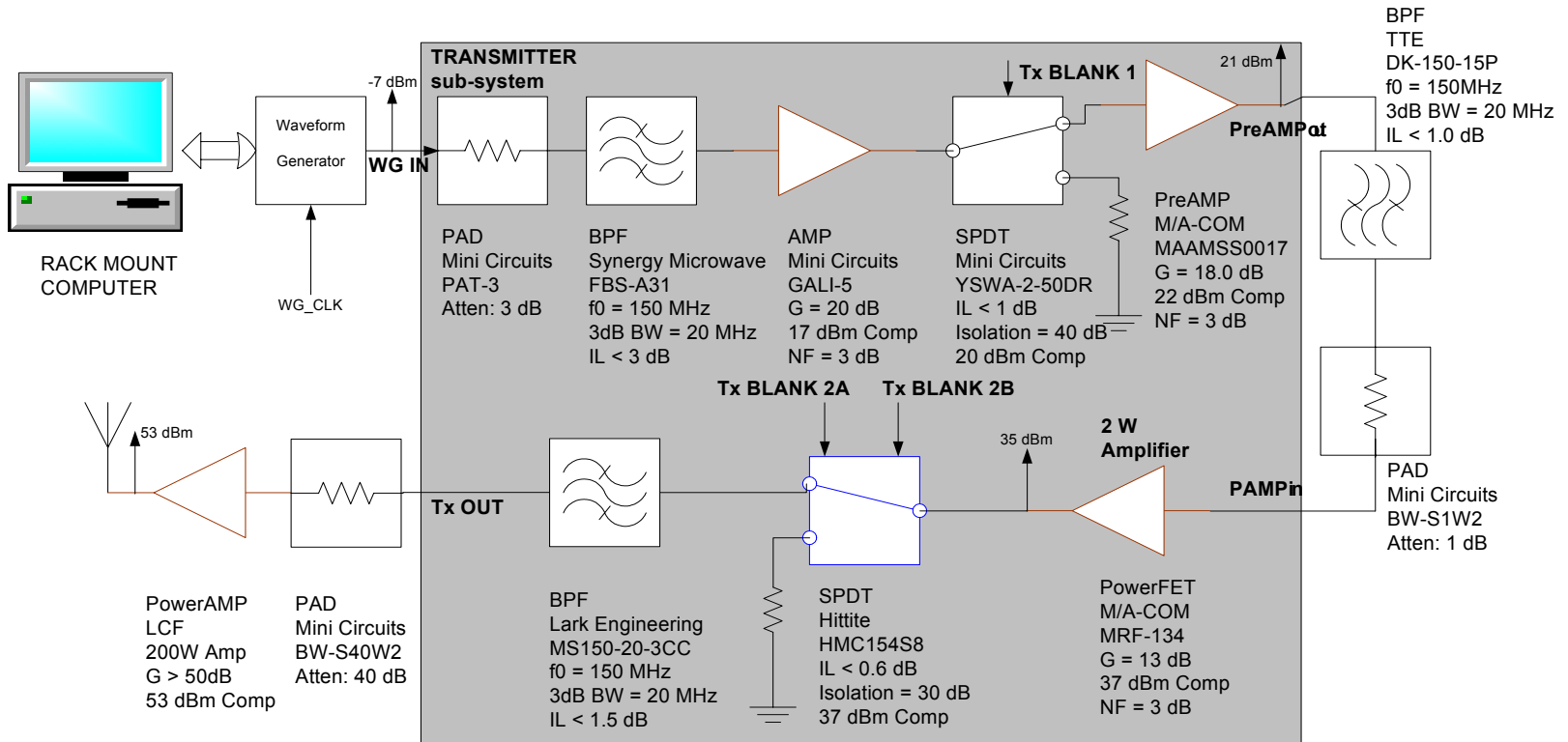
Clock synchronization section design: ACORDS

- Generates clocks for the digital system
 - Trigger section – 10 MHz sinusoid
 - WG - 110 MHz PECL signal
 - DAC – 55 MHz PECL signal
- Whole system synchronized to the base clock
- All surface-mount except LPF





Transmitter RF section design: ACORDS



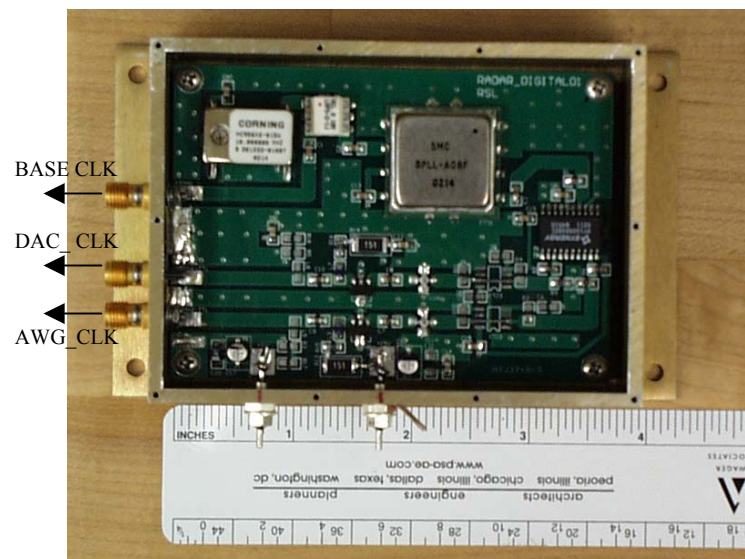
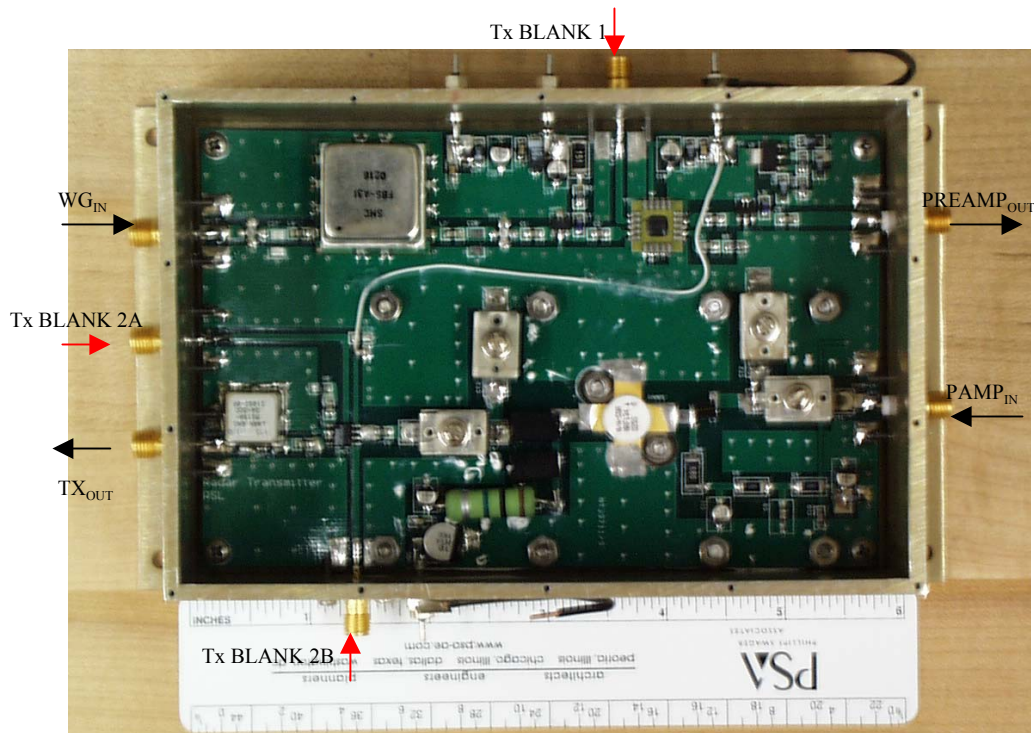


ACORDS Transmitter construction

- Protel 99 for schematic entry and Printed Circuit Board (PCB) layout
- Transmitter sub-system section
 - Designed on a 3.6'' x 5.6'' two layer FR4 board using microwave techniques
 - Aluminum block as heat sink
 - Packaged in a 4'' x 6'' standard RF enclosure
- Clock synchronization section
 - 2.6'' x 3.6'' two layer FR4 board
 - Packaged in a 3'' x 4'' enclosure



ACORDS Transmitter module



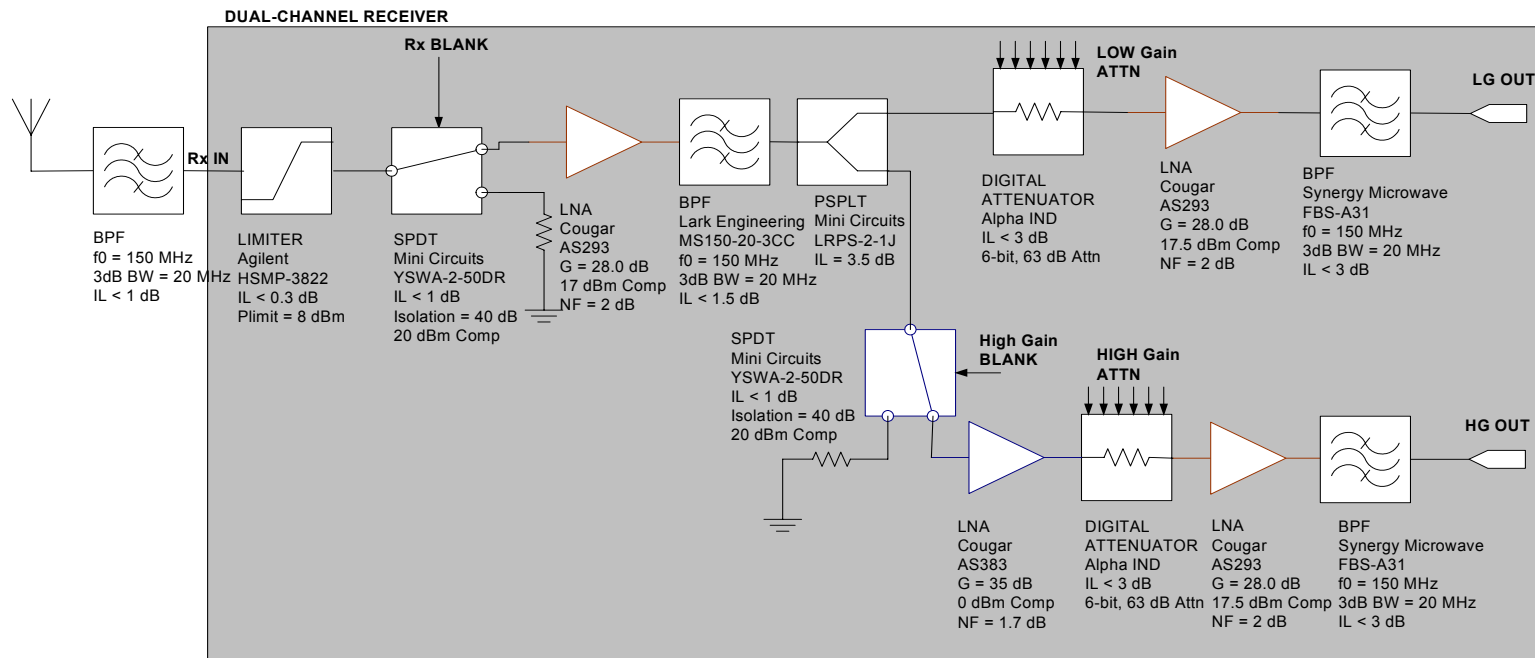


ACORDS dual-channel receiver design

- Consists of RF section only
- LG channel - 44 dB gain, HG channel - 80 dB gain
- Amplifiers from Cougar Components selected
- Two RFICs combined to form 6-bit, 63 dB attenuator
- Gain controlled by a digital attenuator in each channel



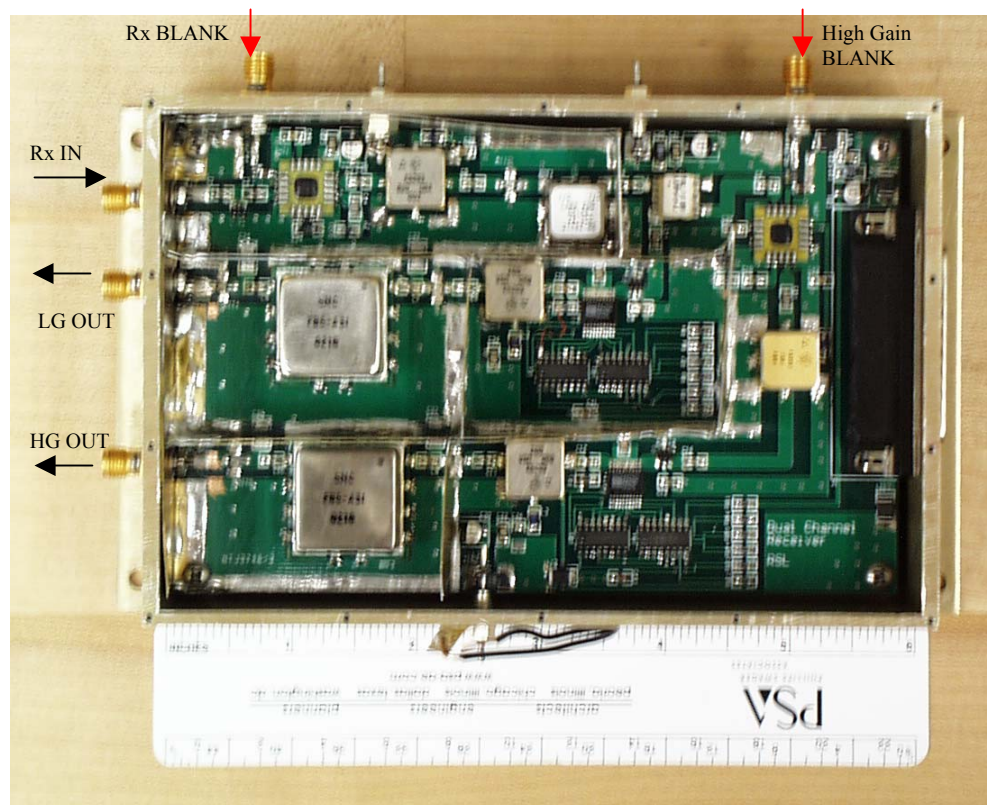
Block diagram of the receiver RF section: ACORDS





ACORDS receiver RF module

- Receiver RF section
 - Designed on a 3.6'' x 5.6'' two layer FR4 board
 - Packaged in a 4'' x 6'' standard RF enclosure
- Internal RF shielding
- 25-pin DB connector





ACORDS digital system

- Developed by Torry Akins
- Three modules
 - 1) Waveform generator
 - 2) Trigger circuit
 - 3) Data acquisition
- Advantages over existing digital system
 - Digital chirp waveform generation
 - Undersampling of LG and HG channel outputs
 - Complete control over radar system
- Universal Serial Bus (USB) interface with host computer



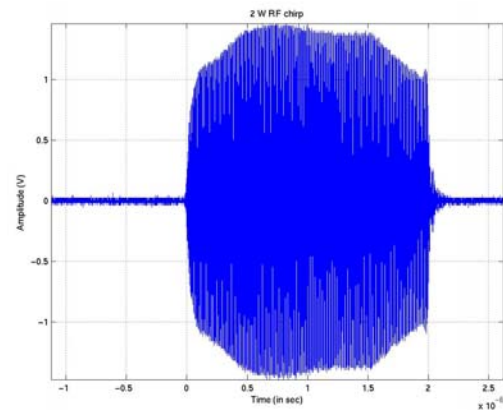
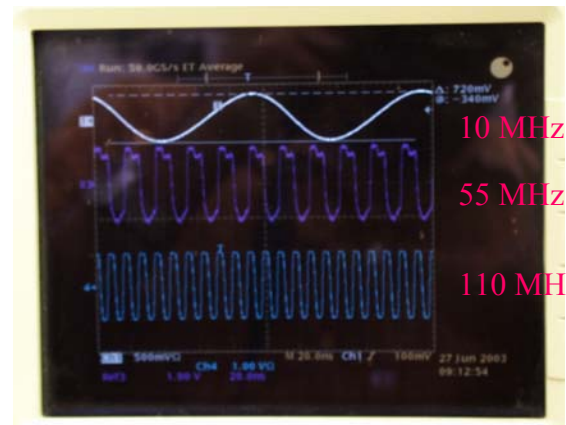
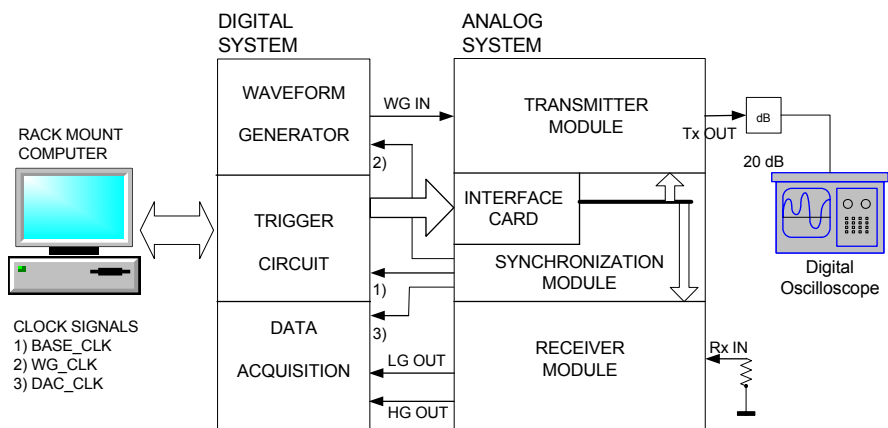
ACORDS analog system

- Linear power supply, individual modules housed in an 18'' x 24'' x 7'' chassis
- Interface card
- Chassis – common electrical ground



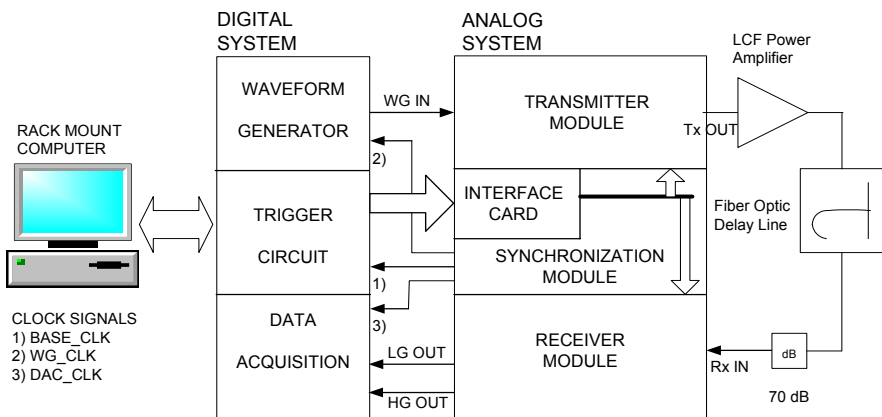


Laboratory testing of the ACORDS system

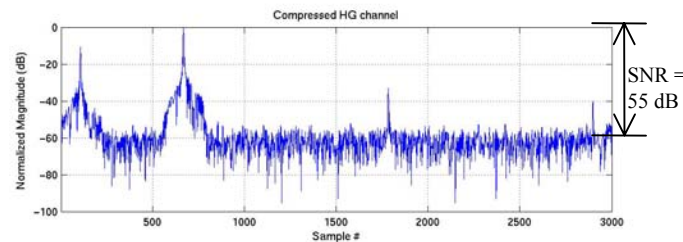
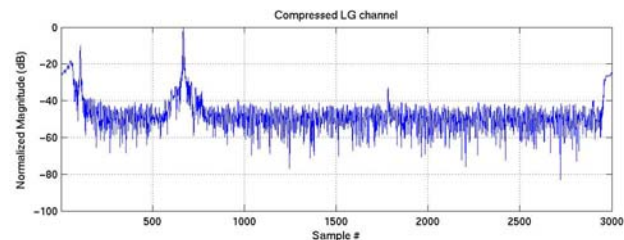
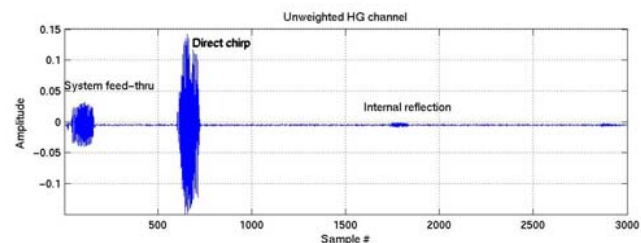
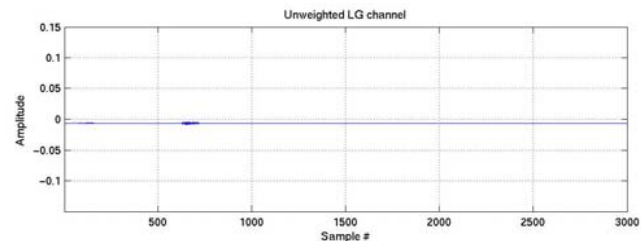




Loop-back test of the ACORDS system



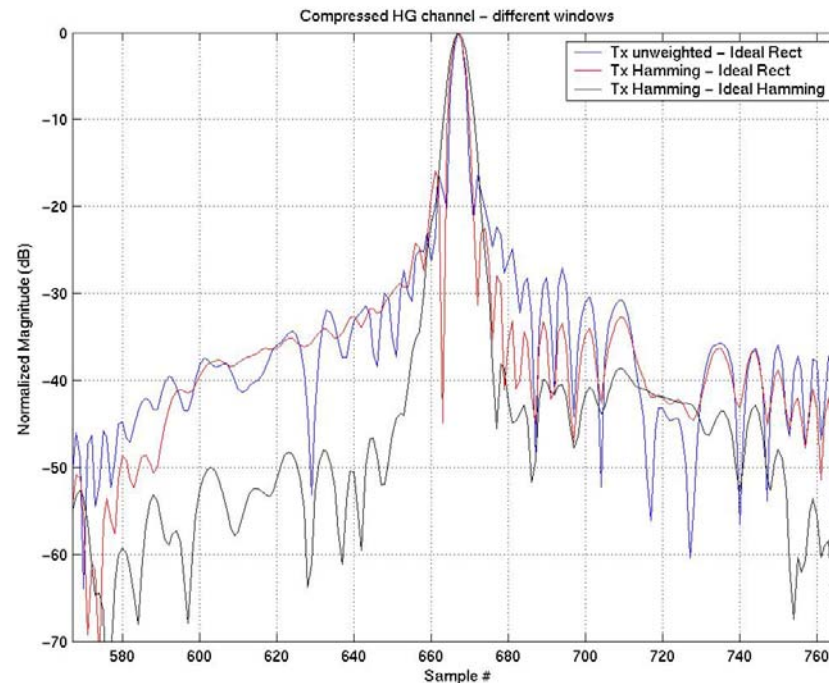
- Practical loop sensitivity (dB)
 - SNR + delay-line attenuation + pad
 - 194 dB
 - Increased by transmitting a long pulse width
- Sidelobe performance
 - 16 dB below main lobe without weighting





Compression results with different windows

- Tested for different windows
- Sidelobe performance
36 dB below main lobe





Field experiment of the ACORDS system

- 2003 field season
- Installed on NASA P3 aircraft
- Common radar settings of ACORDS
- Three different cases
 - Good internal layering
 - Glacier in northern Greenland
 - Glacier in southern Greenland
- Data processing included
 - Pre-integrations



Raw data

- Compensation for gain difference
- Pulse compression
- Post coherent & incoherent averages

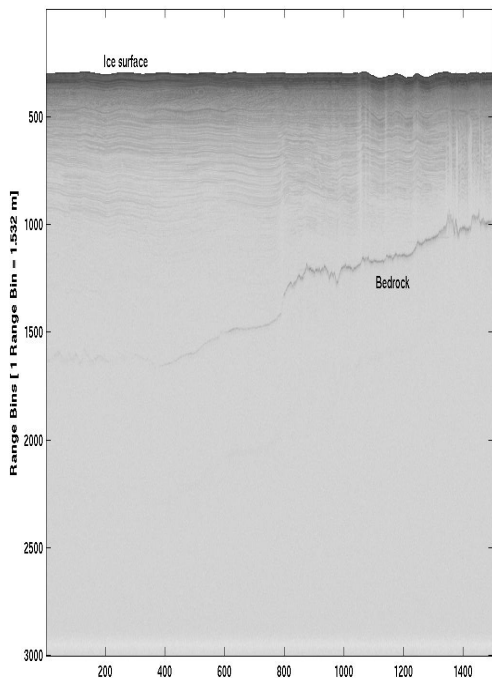
First order processed data

Description	Characteristic	Units
Tx pulse width	3.0	μs
Tx weighting	Hamming	-
LG attenuation	36	dB
HG attenuation	20	dB



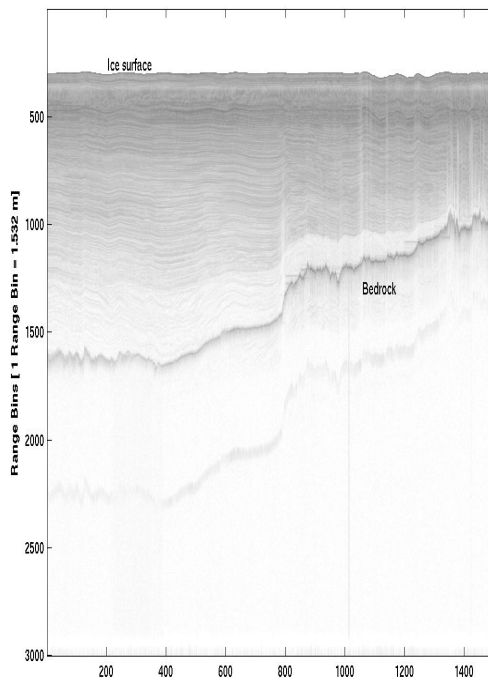
Radio echogram with good layering: Results

LG radio echogram



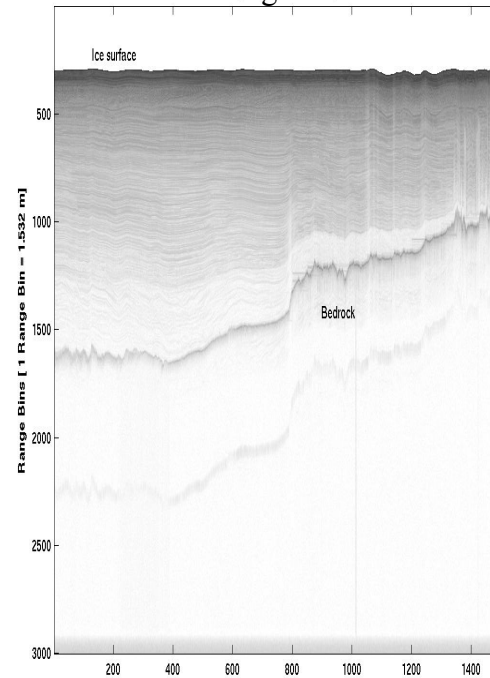
Geolocation								
Latitude:	77.593N	77.523N	77.448N	77.369N	77.286N	77.199N	77.128N	77.052N
Longitude:	55.829W	56.854W	57.879W	58.888W	59.897W	60.897W	61.816W	62.826W
Distance:	0.0 km	23.8 km	51.7 km	77.8 km	104.0 km	130.4 km	158.9 km	183.3 km

HG radio echogram



Geolocation								
Latitude:	77.593N	77.523N	77.448N	77.369N	77.286N	77.199N	77.128N	77.052N
Longitude:	55.829W	56.854W	57.879W	58.888W	59.897W	60.897W	61.816W	62.826W
Distance:	0.0 km	23.8 km	51.7 km	77.8 km	104.0 km	130.4 km	158.9 km	183.3 km

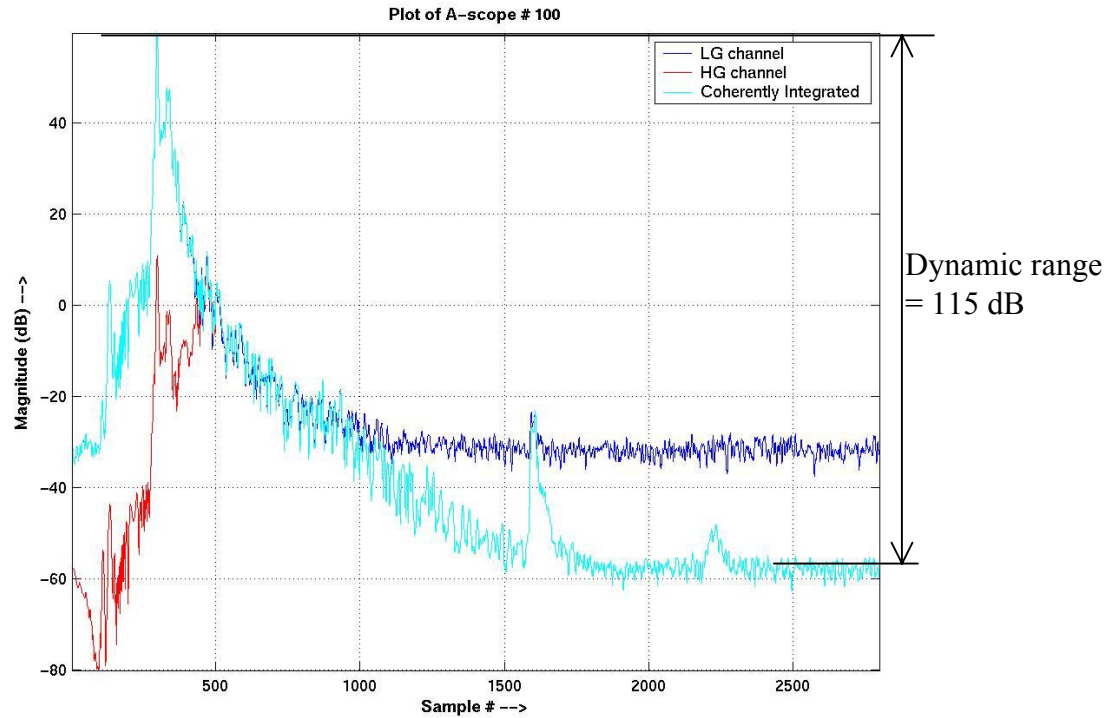
Coherently integrated radio echogram



Geolocation								
Latitude:	77.593N	77.523N	77.448N	77.369N	77.286N	77.199N	77.128N	77.052N
Longitude:	55.829W	56.854W	57.879W	58.888W	59.897W	60.897W	61.816W	62.826W
Distance:	0.0 km	23.8 km	51.7 km	77.8 km	104.0 km	130.4 km	158.9 km	183.3 km



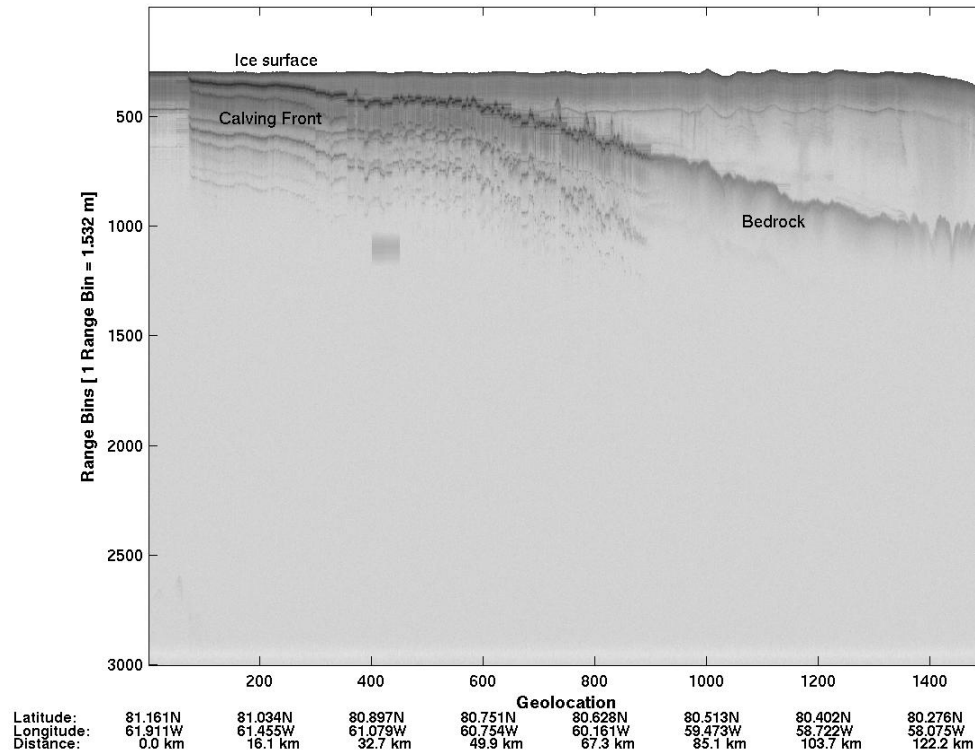
Amplitude-scope comparison: Results





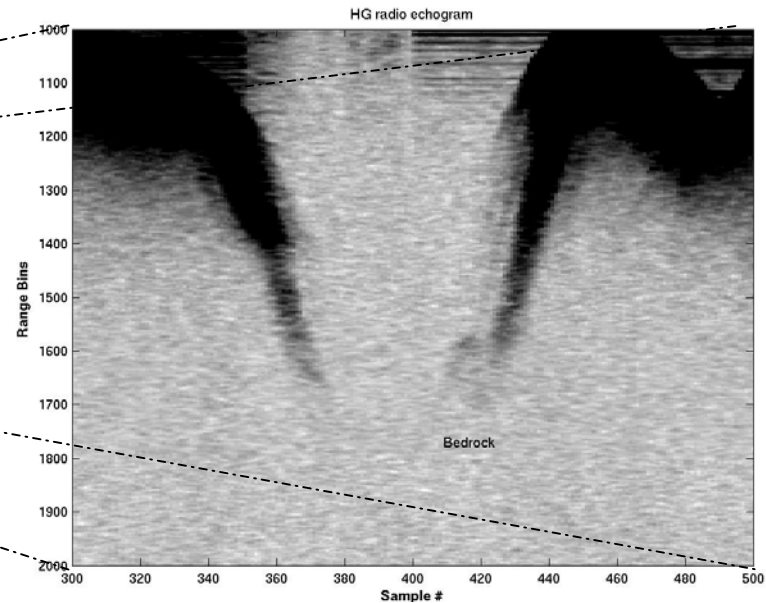
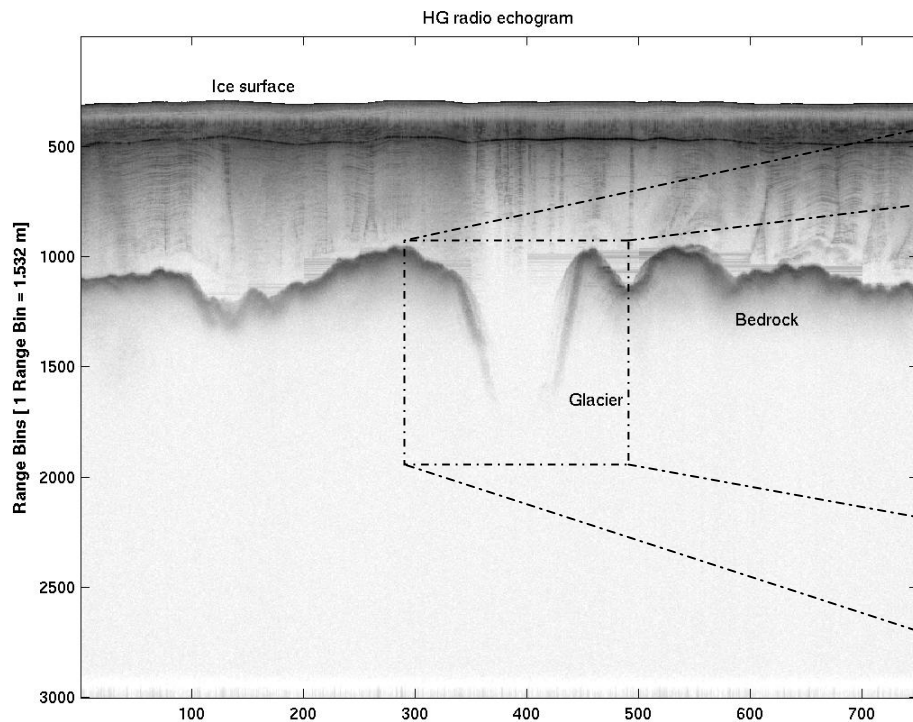
Radio echogram over Petermann Glacier: Results

LG radio echogram of Petermann Glacier





Radio echogram over Jacobshavn Glacier: Results

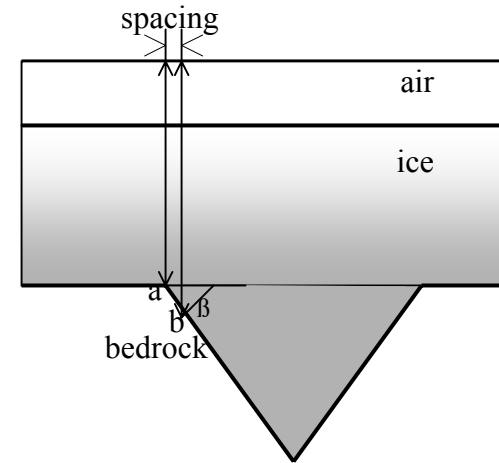


Latitude:	69.392N	69.344N	69.296N	69.251N	69.208N	69.166N	69.125N	69.085N
Longitude:	48.435W	48.439W	48.446W	48.451W	48.455W	48.461W	48.465W	48.47W
Distance:	0.0 km	5.4 km	10.7 km	15.7 km	20.5 km	25.1 km	29.7 km	34.2 km



Summary of results

- LG and HG echograms integrated
- Good control over gain
- Performance over Jacobshavn glacier similar
 - Post-coherent integrations canceling out the bedrock in the channel
 - Two-way phase shift exceeds 45 degrees between adjacent samples
 - Phase compensation needed
 - Post-incoherent integrations insufficient





Conclusions

- Designed and developed
 - Transmitter sub-system
 - Dual-channel receiver
- Integrated in an rack-mount chassis
- Tested with the digital system
- ACORDS performance
 - Reduced sidelobes by 9 dB
 - Improved loop sensitivity
- Easy to operate
- Tested successfully in 2003 field experiment



Future recommendations

- Integrate 200 W power amplifier with transmitter sub-system
- Apply deconvolution techniques to correct for system effects
- Develop an algorithm that takes in to account the slope of the bedrock before coherent integrations
- Reduce number of pre-coherent integrations



Questions?

